New Guaranteed H_{∞} Performance State Estimation for Delayed Neural Networks

Won Il Lee and PooGyeon Park

Abstract—In this paper, a new guaranteed performance state estimation problem for static neural networks with timevarying delay is investigated. A new Lyapunov-Krasovskii functional is introduced to improve the performance. Moreover, with the help of lower bound lemma, an upper-bound of a linear combination of positive functions weighted by the inverses of convex parameters is obtained. Two simulation examples are given to prove the effectiveness of the proposed theorem.

Index Terms—State estimation, static neural networks, H-infinite performance, reciprocally convex approach, Timevarying delay.

I. INTRODUCTION

Neural networks have attracted considerable attention from academic research and industrial applications during the past decades. Various successful applications have been founded in many fields such as pattern recognition, image processing, optimization problems, and adaptive control. Time delay is widely exists in many practical systems such as chemical or process control systems and networked control systems [1]. Also, time delay may exist in neural networks because of their finite switching speeds and communication time. Since these time-delay may induce system instability and performance degradation, the stability analysis of delayed neural networks has become an important issue, and many results have been reported in the literature [2]-[4].

State estimation problem of neural network is very practical and theoretically important issue, which has been studied in recent years [5]-[7]. In many practical applications, the neuron states are not always measurable in the neural networks outputs since it may be very difficult and expensive to acquire all the state information of the neuron states in large-scale neural networks. But the state information may be certainly necessary for some applications such as system modeling and state feedback control. Therefore, in this case, the neuron states should be estimated by measurements, it

Manuscript received August 1, 2012; revised September 3, 2012.

This research was supported by the MKE(The Ministry of Knowledge Economy), Korea, under the ITRC (Information Technology Research Center) support program supervised by the NIPA(National IT Industry Promotion Agency) (NIPA-2012 -H0301-12-2002) & NIPA-2012-(H0301-12-1003)). This research was supported by World Class University program funded by the Ministry of Education, Science and Technology through the National Research Foundation of Korea (R31-10100).

Won II Lee is with the Department of Electrical Engineering, Pohang University of Science and Technology (POSTECH), Pohang, 790-784 Republic of Korea (e-mail: wilee@ postech.ac.kr).

PooGyeon Park is with the Division of ITCE and Department of Electrical Engineering, Pohang University of Science and Technology (POSTECH), Pohang, 790-784 Republic of Korea (e-mail: ppg@postech.ac.kr).

proves the importance of the state estimation problem for neural networks. Recently, [5] proposed a guaranteed performance state estimator for static neural networks with timevarying delay. But in the process of deriving lower bounds of one integral term, [5] introduced an approximation leading to a little conservativeness.

In this paper, we propose a new guaranteed performance state estimator for delayed neural networks based on a new Lyapunov-Krasovskii functional. By applying [8]'s lower bound lemma, an improved performance is obtained.

This paper is organized as follows. The state estimation problem is formulated in Section 2. Section 3 proposes a new guaranteed H-infinity performance state estimator for delayed static neural networks. In Section 4, two simulation examples are given to prove the effectiveness of the proposed theorem.

II. PROBLEM FORMULATION

Consider the delayed static neural network subject to noise disturbances:

$$\Sigma_{w} : \dot{x}(t) = -Ax(t) + f(Wx(t - d(t)) + J) + B_{1}w(t), \quad (1)$$

$$y(t) = Cx(t) + Dx(t - d(t)) + B_2 w(t),$$
(2)

$$z(t) = Hx(t), \tag{3}$$

$$x(s) = \phi(s), \quad s \in [-\tau, 0],$$
 (4)

where $x(t) = [x_1(t), x_2(t), \dots, x_n(t)]^T \in \mathbf{R}^n$ is the neuron state vector; $y(t) \in \mathbf{R}^m$ is the network output measurement; $z(t) \in \mathbf{R}^p$ is a linear combination of the states to be estimated; $w(t) \in \mathbf{R}^q$ is a noise disturbance belonging to $L_2[0,\infty)$; $A = diag\{a_1, a_2, \dots, a_n\}$ is a diagonal matrix with positive entries $a_i > 0$; $W^{n \times n}$ is the interconnection matrices representing the weighting coefficients of the neurons; $f(x(t)) = [f_1(x_1(t), f_2(x_2(t)), \dots, f_n(x_n(t))]^T$ is the neuron activation function; $J = [J_1, J_2, \dots, J_n]^T$ is an external input vector; $\phi(s)$ is an initial condition on $[-\tau, 0]$, and d(t) is a time-varying delay of the system satisfying

$$0 \le d(t) \le \tau, \quad 0 \le \dot{d}(t) \le \mu . \tag{5}$$

In this paper, we choose a state estimator for the neural network (Σ_{N}) as

$$\Sigma_{s} : \hat{x}(t) = -A\hat{x}(t) + f(W\hat{x}(t-d(t)) + J) + K[y(t) - C\hat{x}(t) - D\hat{x}(t-d(t))].$$
(6)

$$\hat{z}(t) = H\hat{x}(t),$$
 (7)

$$\hat{x}(s) = 0, \ s \in [-\tau, 0],$$
 (8)

where $\hat{x}(t) \in \mathbf{R}^n$, $\hat{z}(t) \in \mathbf{R}^p$, and *K* is the state estimator gain matrix to be determined.

Define the errors to be $e(t) = x(t) - \hat{x}(t)$ and $\overline{z}(t) = z(t) - \hat{z}(t)$. Then the error-state system is represented by

$$\Sigma_E : \dot{e}(t) = -(A + KC)e(t) - KDe(t - d(t)) + \psi(We(t - d(t)),$$
$$\hat{c}(t - d(t))) + (B - KB)w(t)$$
(0)

$$\kappa(t - d(t))) + (B_1 - KB_2)w(t),$$
(9)

$$\overline{z}(t) = He(t) \tag{10}$$

where $\psi(We(t), \hat{x}(t)) = f(Wx(t) + J) - f(W\hat{x}(t) + J)$.

Definition 1: The error system (Σ_{E}) is said to be globally stable with H_{∞} performance γ if, for some scalar $\gamma > 0$, there exists a proper state estimator (Σ_{s}) such that the equilibrium point of the resulting error system (9) with $w(t) \equiv 0$ is globally asymptotically stable, and

$$\left\|\overline{z}(t)\right\|_{2} < \gamma \left\|w(t)\right\|_{2} \tag{11}$$

Under zero-initial conditions for all nonzero $w(t) \in L_2[0,\infty)$, where $\|\eta(t)\|_2 = \sqrt{\int_0^\infty \eta^T(t)\eta(t)dt}$.

Assumption 1. The neuron activation functions in (1), $f_i(\cdot)$, satisfy the following Lipschitz condition

$$0 \le \frac{f_i(x) - f_i(y)}{x - y} \le l_i, \ x \ne y \in \mathbf{R}, \ (i=1,2,\cdots,n)$$
(12)

with $L = diag(l_1, l_2, \dots, l_n) > 0$.

III. GUARANTEED H_{ω} PERFORMANCE STATE ESTIMATOR

This section is dedicated to the design of a guaranteed H_{∞} performance state estimator for the delayed static neural network. A delay dependent LMI based condition will be established.

Theorem 1. Under Assumption 1, given positive scalars τ , μ , and prescribed constant $\gamma > 0$, state estimation problem of the delayed static neural network (Σ_N) with guaranteed H_{∞} performance is solvable if there exist appropriately dimensioned matrices P > 0, $Q_i > 0$ (i=1,2,3), $S_i > 0$,

 $R_{j} > 0$ (j=1,2), Z and two diagonal matrices $\Lambda = diag(\lambda_{1}, \lambda_{2}, \dots, \lambda_{n}) > 0$, $\Gamma = diag(\gamma_{1}, \gamma_{2}, \dots, \gamma_{n}) > 0$ such that the following LMIs hold

$$\begin{bmatrix} S_{2} + R_{1} & S \\ S & S_{2} + R_{2} \end{bmatrix} > 0,$$
(13)





where

$$\begin{split} \overline{\Omega}_{11} &= -PA - A^{T}P - GC - C^{T}G^{T} + Q_{1} + Q_{2} \\ &- S_{2} + H^{T}H + \tau S_{1} - 2R_{1}, \\ \overline{\Omega}_{12} &= -GD + S_{2} - S, \quad \overline{\Omega}_{19} = PB_{1} - GB_{2}, \\ \overline{\Omega}_{110} &= -\tau A^{T}P - \tau C^{T}G^{T}, \\ \Omega_{22} &= -(1 - \mu)Q_{1} - 2S_{2} + S + S^{T} - 2R_{2} - 2R_{1}, \\ \Omega_{55} &= \Omega_{66} = -2R_{1} - 2R_{2} - \tau S_{1}, \quad \Omega_{88} = -(1 - \mu)Z - 2\Gamma, \\ \overline{\Omega}_{910} &= \tau B_{1}^{T}P - \tau B_{2}^{T}G^{T}. \end{split}$$

In this case, a desired the state estimator gain matrix K is given as $K = P^{-1}G$.

Proof. Choose a Lyapunov-Krasovskii functional candidate as

$$V(e(t)) = e^{T}(t)Pe(t) + \int_{t-d(t)}^{t} e^{T}(s)Q_{1}e(s)ds + \int_{t-\tau}^{t} e^{T}(s)Q_{2}e(s) + \int_{t-\tau}^{t} \dot{e}^{T}(s)Q_{3}\dot{e}(s)ds + \int_{-\tau}^{0} \int_{t+\theta}^{t} e^{T}(s)S_{1}e(s)dsd\theta + \tau \int_{-\tau}^{0} \int_{-\tau}^{0} \int_{t+\theta}^{t} \dot{e}^{T}(s)S_{2}\dot{e}(s)dsd\theta + \int_{-\tau}^{0} \int_{\eta}^{0} \int_{t+\theta}^{t} \dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta d\eta + \int_{-\tau}^{0} \int_{-\tau}^{\eta} \int_{-\tau}^{t} \int_{t+\theta}^{t} \dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta d\eta + \int_{t-\tau}^{0} \int_{-\tau}^{\eta} \int_{t+\theta}^{t} \dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta d\eta + \int_{t-\tau}^{0} \int_{t-\theta}^{\eta} (We(s), \hat{x}(s))Z\psi(We(s), \hat{x}(s))ds.$$
(16)

Calculating the time-derivative of V(e(t)) along the traject-

ories of system (1) and noting that d(t) satisfies (5), it yields

$$\begin{split} \dot{V}(e(t)) &\leq e^{T}(t)[-P(A+KC) - (A+KC)^{T}P + Q_{1} + Q_{2} + \tau S_{1}]e(t) \\ &- 2e^{T}(t)PKDe(t-d(t)) + 2e^{T}(t)P\psi(We(t-d(t)), \hat{x}(t-d(t))) \\ &+ 2e^{T}(t)P(B_{1} - KB_{2})w(t) - (1-\mu)e^{T}(t-d(t))Q_{2}e(t-d(t)) \\ &+ \psi^{T}(We(t), \hat{x}(t))Z\psi(We(t), \hat{x}(t)) - (1-\mu)\psi^{T}(We(t-d(t))), \\ &\hat{x}(t-d(t)))Z\psi(We(t-d(t)), \hat{x}(t-d(t))) - e^{T}(t-\tau)Q_{2}e(t-\tau) \\ &- \dot{e}^{T}(t-\tau)Q_{3}\dot{e}(t-\tau) + \dot{e}^{T}(t)[S_{2} + Q_{3} + \frac{\tau^{2}}{2}(R_{1} + R_{2})]\dot{e}(t) \\ &- \int_{t-\tau}^{t}e^{T}(s)S_{1}e(s)ds - \tau \int_{t-\tau}^{t}\dot{e}^{T}(s)S_{2}\dot{e}(s)ds \\ &- \int_{-\tau}^{0}\int_{t+\theta}^{t}\dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta - \int_{-\tau}^{0}\int_{t-\tau}^{t+\theta}\dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta. \end{split}$$

$$(17)$$

Using Jensen's inequality [9], one can obtain

$$-\int_{t-\tau}^{t} e^{T}(s)S_{1}e(s)ds$$

$$= -\int_{t-\tau}^{t-d(t)} e^{T}(s)S_{1}e(s)ds - \int_{t-d(t)}^{t} e^{T}(s)S_{1}e(s)ds$$

$$\leq -(\tau - d(t))[\frac{1}{\tau - d(t)}\int_{t-\tau}^{t-d(t)} e^{T}(s)ds]S_{1}[\frac{1}{\tau - d(t)}\int_{t-\tau}^{t-d(t)} e(s)ds]$$

$$- d(t)[\frac{1}{d(t)}\int_{t-d(t)}^{t} e^{T}(s)ds]S_{1}[\frac{1}{d(t)}\int_{t-d(t)}^{t} e(s)ds],$$
(18)

$$-\tau \int_{t-\tau}^{t} \dot{e}^{T}(s) S_{2} \dot{e}(s) ds$$

$$= -\tau \int_{t-\tau}^{t-d(t)} \dot{e}^{T}(s) S_{2} \dot{e}(s) ds - \tau \int_{t-d(t)}^{t} \dot{e}^{T}(s) S_{2} \dot{e}(s) ds$$

$$\leq -\frac{\tau}{\tau - d(t)} [e^{T}(t - d(t)) - e^{T}(t - \tau)] S_{2} [e(t - d(t)) - e(t - \tau)]$$

$$-\frac{\tau}{d(t)} [e^{T}(t) - e^{T}(t - d(t))] S_{2} [e(t) - e(t - d(t)],$$
(19)

$$\begin{aligned} &-\int_{-\tau}^{0}\int_{t+\theta}^{t}\dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta \\ &= -\int_{-d(t)}^{0}\int_{t+\theta}^{t}\dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta - \int_{-\tau}^{-d(t)}\int_{t+\theta}^{t-d(t)}\dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta \\ &-\int_{-\tau}^{-d(t)}\int_{t-d(t)}^{t}\dot{e}^{T}(s)R_{1}\dot{e}(s)dsd\theta \\ &\leq -2[e^{T}(t) - \frac{1}{d(t)}\int_{t-d(t)}^{t}e^{T}(s)ds]R_{1}[e(t) - \frac{1}{d(t)}\int_{t-d(t)}^{t}e(s)ds] \\ &- 2[e^{T}(t-d(t)) - \frac{1}{\tau-d(t)}\int_{t-\tau}^{t-d(t)}e^{T}(s)ds]R_{1}[e(t-d(t)) \\ &-\frac{1}{\tau-d(t)}\int_{t-\tau}^{t-d(t)}e(s)ds] + [e^{T}(t) - e^{T}(t-d(t))]R_{1}[e(t) \\ &-e(t-d(t))] - \frac{\tau}{d(t)}[e^{T}(t) - e^{T}(t-d(t))]R_{1}[e(t) - e(t-d(t))], \end{aligned}$$

$$\begin{aligned} &-\int_{-\tau}^{0}\int_{t-\tau}^{t+\theta}\dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta \\ &= -\int_{-d(t)}^{0}\int_{t-d(t)}^{t+\theta}\dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta - \int_{-\tau}^{-d(t)}\int_{t-\tau}^{t+\theta}\dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta \\ &-\int_{-d(t)}^{0}\int_{t-\tau}^{t-d(t)}\dot{e}^{T}(s)R_{2}\dot{e}(s)dsd\theta \\ &\leq -2[\frac{1}{d(t)}\int_{t-\tau}^{t}e^{T}(s)ds - e^{T}(t-d(t))]R_{2}[\frac{1}{d(t)}\int_{t-d(t)}^{t}e(s)ds \\ &-e(t-d(t))] - 2[\frac{1}{\tau-d(t)}\int_{t-\tau}^{t-d(t)}e^{T}(s)ds - e^{T}(t-\tau)] \\ &\times R_{2}[\frac{1}{\tau-d(t)}\int_{t-\tau}^{t-d(t)}e(s)ds - e(t-\tau)] \\ &+ [e^{T}(t-d(t)) - e^{T}(t-\tau)]R_{2}[e(t-d(t)) - e(t-\tau)] \\ &-\frac{\tau}{\tau-d(t)}[e^{T}(t-d(t)) - e^{T}(t-\tau)]R_{2}[e(t-d(t)) - e(t-\tau)]. \end{aligned}$$

Let us define $\alpha = d(t) / \tau$ and

$$\lambda(t) = col\{\sqrt{\frac{1-\alpha}{\alpha}}(e(t) - e(t - d(t)), \sqrt{\frac{\alpha}{1-\alpha}}(e(t - d(t)) - e(t - \tau))\}$$

Then, gathering the positive functions weighted by the inverses of the convex parameters $\{\alpha, 1-\alpha\}$, or equivalently $\{\tau / d(t), \tau / (\tau - d(t))\}$ from (19)-(21) and applying lower bound lemma [8] for *S* satisfying (13), we have

$$\lambda^{T}(t) \begin{bmatrix} S_{2} + R_{1} & S \\ S^{T} & S_{2} + R_{2} \end{bmatrix} \lambda(t) \geq 0$$

which produces an upper-bound as

$$-\frac{\tau}{d(t)} [e^{\tau}(t) - e^{\tau}(t - d(t))](S_{2} + R_{1})[e(t) - e(t - d(t))]$$

$$-\frac{\tau}{\tau - d(t)} [e^{\tau}(t - d(t)) - e^{\tau}(t - \tau)](S_{2} + R_{2})[e(t - d(t)) - e(t - \tau)]$$

$$\leq -\left[\frac{e(t) - e(t - d(t))}{e(t - d(t)) - e(t - \tau)}\right]^{\tau} \left[\frac{S_{2} + R_{1}}{S^{\tau}} \frac{S}{S_{2} + R_{2}}\right] \left[\frac{e(t) - e(t - d(t))}{e(t - d(t)) - e(t - \tau)}\right].$$
(22)

By assumption 1, one can obtain that, for any $W_i e \neq 0$,

$$0 \leq \frac{\psi_i(W_i e, \hat{x})}{W_i e} = \frac{f_i(W_i x + J_i) - f_i(W_i \hat{x} + J_i)}{W_i x - W_i \hat{x}} \leq l_i,$$

where $W_i = [w_{i1}, w_{i2}, \dots, w_{in}]$ is the i-th row vector of W. Then the following inequalities hold,

$$0 \leq -2\psi^{T} (We(t), \hat{x}(t)) \Lambda \psi (We(t), \hat{x}(t)) + 2\psi^{T} (We(t), \hat{x}(t)) \Lambda LWe(t),$$
(23)

925

$$0 \leq -2\psi^{T} (We(t - d(t)), \hat{x}(t - d(t))) \Gamma \psi (We(t - d(t)), \hat{x}(t - d(t))) + 2\psi^{T} (We(t - d(t)), \hat{x}(t - d(t))) \Gamma L We(t - d(t)),$$
(24)

For any diagonal matrices $\Lambda = diag(\lambda_1, \lambda_2, \dots, \lambda_n) > 0$ and $\Gamma = diag(\gamma_1, \gamma_2, \dots, \gamma_n) > 0$.

Define
$$J(t) = \int_0^t [\overline{z}^T(s)\overline{z}(s) - \gamma^2 w^T(s)w(s)]ds$$
 for $t > 0$.

Then, for any non-zero $w(t) \in L_2[0,\infty)$,

$$J(t) \leq \int_{0}^{t} [\overline{z}^{T}(s)\overline{z}(s) - \gamma^{2}w^{T}(s)w(s)]ds + Ve(t)) - V(e(0))$$

$$= \int_{0}^{t} [\overline{z}^{T}(s)\overline{z}(s) - \gamma^{2}w^{T}(s)w(s) + \dot{V}(e(s))]ds.$$
 (25)

From the condition (10) and (17)- (24), it can be seen that

$$\overline{z}^{T}(t)\overline{z}(t) - \gamma^{2}w^{T}(t)w(t) + \dot{V}(e(t))$$

$$\leq \xi^{T}(t)[\Omega_{1} + \tau^{2}\Omega_{2}^{T}X\Omega_{2} - d(t)\Omega_{3}^{T}S_{1}\Omega_{3} - (\tau - d(t))\Omega_{4}^{T}S_{1}\Omega_{4}]\xi(t),$$
(26)

where

$$\xi(t) = [e^{\tau}(t), e^{\tau}(t-d(t)), e^{\tau}(t-\tau), \dot{e}^{\tau}(t-\tau), \\ \frac{1}{d(t)} \int_{t-d(t)}^{t} e(s) ds, \frac{1}{\tau - d(t)} \int_{t-\tau}^{t-d(t)} e(s) ds, \psi^{\tau}(We(t), \hat{x}(t)), \\ \psi^{\tau}(We(t-d(t)), \hat{x}(t-d(t))), w^{\tau}(t)]^{\tau},$$
(27)

Since $-d(t)\Omega_3^T S_1\Omega_3 - (\tau - d(t))\Omega_4^T S_1\Omega_4$ is a convex combination of the matrices $\Omega_3^T S_1\Omega_3$ and $\Omega_4^T S_1\Omega_4$ on d(t), it can be non-conservatively handled by two boundary LMIs: one for d(t) = 0 and the other for $d(t) = \tau$. Pre- and postmultiplying two LMIs by $diag\{I, I, I, I, I, I, I, I, I, PX^{-1}\}$. Using following inequality for any real P > 0 and X > 0

$$PX^{-1}P - 2P + X = (P - X)X^{-1}(P - X) \ge 0,$$

$$-PX^{-1}P \le -2P + X,$$
(29)

And applying the change of variable such that $K = P^{-1}G$, one can deduce the LMIs (13)-(15) imply $\overline{z}^{T}(s)\overline{z}(s) - \gamma^{2}w^{T}(s) \times w(s) + \dot{V}(e(s)) < 0$ for $w(t) \neq 0$. Therefore, J(t) < 0 from (25) for t > 0, thus (11) holds.

Globally asymptotically stability of the equilibrium point of the error system (9) with $w(t) \equiv 0$ is achieved if $\dot{V}(e(t)) < 0$ holds. One can easily prove that the condition $\dot{V}(e(t)) < 0$ is guaranteed by the LMIs (13)-(15). We skip the specific proof due to space limitation.

IV. SIMULATION EXAMPLES

Two simulation examples are given in this section to illustrate the effectiveness of the developed approach.

Example 1. Consider a delayed static neural network (Σ_N) with the following parameters:

$$A = \begin{bmatrix} 0.96 & 0 & 0 \\ 0 & 0.8 & 0 \\ 0 & 0 & 1.48 \end{bmatrix}, W = \begin{bmatrix} 0.5 & 0.3 & -0.36 \\ 0.1 & 0.12 & 0.5 \\ -0.42 & 0.78 & 0.9 \end{bmatrix},$$
$$B = \begin{bmatrix} 0.1 \\ 0.2 \\ 0.1 \end{bmatrix}, B_2 = -0.1, C = \begin{bmatrix} 1 & 0 & -2 \end{bmatrix},$$
$$D = \begin{bmatrix} 0.5 & 0 & -1 \end{bmatrix}, H = \begin{bmatrix} 1 & 1 & 0 \\ 1 & 0 & -1 \\ 0 & 1 & 1 \end{bmatrix}, J = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}.$$

The activation function, the time varying delay, and the noise disturbance are taken as $f(x) = \tanh(x)$ with L = I, $d(t) = 0.5 + 0.5 \cos(2.4t)$ with $\tau = 1$ and $\mu = 1.2$, and w(t) = 1/(0.8 + 1.2t) for t > 0, respectively.

Then, solving theorem 1 by resorting to the LMI solver in the Matlab LMI Control Toolbox, the state estimator gain matrix can be found as

$$K = \begin{bmatrix} 0.0728 \\ -0.1987 \\ -0.2965 \end{bmatrix},$$

With the optimal H_{∞} performance index $\gamma_{min} = 1.3705$. It is easy to notice that this result is an improved result than $\gamma_{min} = 1.6002$ in [5]. Fig. 1 represents the error e(t) for 10 random initial values.

Example 2. Consider a delayed static neural network (Σ_N) with the following parameters:

$$A = \begin{bmatrix} 1.56 & 0 & 0 \\ 0 & 2.42 & 0 \\ 0 & 0 & 1.88 \end{bmatrix}, W = \begin{bmatrix} -0.3 & 0.8 & -1.36 \\ 1.1 & 0.4 & -0.5 \\ 0.42 & 0 & -0.95 \end{bmatrix},$$

$$B = \begin{bmatrix} 0.2 \\ 0.2 \\ 0.2 \\ 0.2 \end{bmatrix}, B_2 = 0.4, C = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix},$$



Fig. 1. The error e(t) for 10 random initial value.

TABLE I: COMPARISON OF THE OPTIMAL H_{∞} PERFORMANCE INDEX γ

| (τ,μ) | (0.9, 0.3) | (0.9, 0.5) | (0.8, 0.7) |
|--------------|------------|------------|------------|
| [5] | 0.3404 | 0.3871 | 0.2883 |
| Theorem 1 | 0.3247 | 0.3689 | 0.2430 |

$$D = \begin{bmatrix} 2 & 0 & 0 \end{bmatrix}, \ H = \begin{bmatrix} 1 & 0 & 0.5 \\ 1 & 0 & 1 \\ 0 & -1 & 1 \end{bmatrix}, \ L = I.$$

Using Theorem 1, the optimal H_{∞} performance index γ_{\min} can be derived at different (τ, μ) , and these are summarized in Table 1. It is easy to see that our method can obtain much better γ_{\min} than those in [5].

V. CONCLUSION

In this paper, the guaranteed H-infinite performance state estimation problems is studied for delayed static neural networks. Based on a new Lyapunov-Krasovskii functional, we solved the guaranteed H-infinite performance state estimation problem. Moreover, with the help of [8]'s lower bound lemma, we could obtain an improved H-infinite performance results for delayed static neural networks. It is shown that the guaranteed H-infinite performance state estimator gain matrix can be found by solving LMIs. Two simulation examples proved the improvement of the proposed theorem compared to existing one. It is worth noticing that the proposed theorem can be widely applicable in control fields such as state feedback control problems and large scale neural networks, and so on.

REFERENCES

- J. P. Richard, "Time-delay systems: An overview of some recent advances and open problems," *Automatica*, vol. 39, no. 10, pp. 1667-1694, 2003.
- [2] Y. He, G. P. Liu, D. Rees, and M. Wu, "Stability analysis for neural networks with time-varying delay," *IEEE Trans. Neural Networks*, vol. 38, no. 3, pp. 1152-1156, 2008.
- [3] X, M. Zhang and Q. L. Han, "New Lyapunov-Krasovskii functionals for global asymptotic stability of delayed neural networks," *IEEE Trans. Neural Networks*, vol. 20, no. 3, pp. 533-539, 2009.
- [4] Y. Liu, Z. Wang, and X. Liu, "Asymptotic stability for neural networks with mixed time-delays: the discrete-time case," *Neural Networks*, vol. 22, no. 1, pp. 67-74, 2009.
- [5] H. Huang, G. Feng, and J. Cao, "Guaranteed performance state estimation of static neural networks with time-varying delay," *Neurocomputing*, vol. 74, no. 4, pp. 606-616, 2011.
- [6] H. Huang and G. Feng, "Robust state estimation for uncertain neural networks with time-varying delay," *IEEE Trans. Neural Networks*, vol. 19, no. 8, pp. 1329-1339, 2008.
- [7] Y. He, Q. G. Wang, M. Wu and C. Lin, "Delay-dependent state estimation for delayed neural networks," *IEEE Trans. on neural networks*, vol. 17, no. 4, pp. 1077-1081, 2006.
- [8] P. Park, J. W. Ko and C. Jeong, "Reciprocally convex approach to stability of systems with time-varying delays," *Automatica*, vol. 47, no. 1, pp. 235-238, 2011.
- [9] S. Boyd, L. E. Ghaoui, E. Feron, V. Balakrishnan, Linear Matrix Inequalities in System and Control Theory, Philadelphia, PA: SIAM, 1994, pp. 32-33.



Won II Lee received his B.S. degree in electronic and electrical engineering from Kyungpook National University in 2010. He is currently studying toward his Ph.D. at Pohang University of Science and Technology (POSTECH). His current research interests include robust, Linear Parameter Varying (LPV), delayed systems, and Neural Networks.



PooGyeon Park received his B.S. and M.S. degrees in Control and Instrumentation Engineering from Seoul National University, Korea, in 1988 and 1990, respectively, and the Ph.D. degree in Electrical Engineering from Stanford University, U.S.A., in 1995. Since 1996, he has been affiliated with the Division of Electrical and Computer Engineering at Pohang University of Science and Technology, where he is currently a Professor. His current research interests include robust, Linear Parameter

Varying (LPV), Receding Horizon Control (RHC), intelligent, and network-related control theories, signal processing, and wireless communications for personal area network (PAN).